

A DISCRETE FRACTIONAL GABOR EXPANSION FOR TIME-- FREQUENCY SIGNAL ANALYSIS

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ABSTRACT

In this work, we present a discrete fractional Gabor representation on a general, non-rectangular time-frequency lattice. The traditional Gabor expansion represents a signal in terms of time and frequency shifted basis functions, called Gabor logons. This constant-bandwidth analysis uses a fixed, and rectangular time-frequency plane tiling. Many of the practical signals require a more flexible, non-rectangular time-frequency lattice for a compact representation. The proposed fractional Gabor method uses a set of basis functions that are related to the fractional Fourier basis and generate a non-rectangular tiling. Simulation results are presented to illustrate the performance of our method.

Keywords: Spread Time-frequency analysis, Gabor expansion Fractional Fourier Transform.

1. INTRODUCTION

Time-frequency (TF) analysis provides a characterization of signals in terms of joint time and frequency content [1]. One of the fundamental issues in the TF analysis is obtaining the distribution of signal energy over joint TF plane with a delta function concentration [1]. The discrete Gabor expansion is a TF signal decomposition which represents a signal in terms of time and frequency translated basis functions called TF atoms [2],[3]. Gabor basis functions $g_{m,k}(n)$ are obtained by shifting and modulating with a sinusoid a single window function $g(n)$, which results in a fixed and

rectangular TF plane tiling. However, if the signal to be represented is not modeled well by this constant-bandwidth analysis, its Gabor representation displays poor TF localization [4],[5],[6]. Many of the practical signals such as speech, music, biological, and seismic signals have time-varying frequency nature that is not appropriate for sinusoidal analysis [4],[6]. Thus

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the traditional Gabor expansion of such signals will require large number of coefficients yielding a poor TF localization. The compactness of the Gabor representation is improved if the basis functions match the time-varying frequency behavior of the signal [6],[7],[8]. Here we present a new, fractional Gabor expansion that uses a more flexible, non-rectangular TF lattice. The basis functions of the proposed expansion are related to the fractional Fourier basis.

2. THE DISCRETE GABOR EXPANSION

The traditional Gabor expansion [2],[3] represents a signal in terms of time and frequency shifted basis functions, and has been used in various applications to analyze the time-varying frequency content of a signal [9]. Basis functions of the Gabor representation are obtained by translating and modulating with sinusoids a single window function. The discrete Gabor expansion of a finite-support signal $x(n), 0 \leq n \leq N-1$ is given by [3]

$$x(n) = \sum_{m=0}^{M-1} \sum_{k=0}^{K-1} c_{m,k} \tilde{g}_{m,k}(n) \quad (1)$$

where the basis function

$$\tilde{g}_{m,k}(n) = \tilde{g}(n - mL) e^{jw_k n} \quad (2)$$

and $w_k = 2\pi k L' / N$. The Gabor expansion parameters M, K, L , and L' are positive integers constrained by $ML = KL' = N$ where M and K are the number of analysis samples in time and frequency, respectively, and L and L' are the time and frequency steps, respectively. Existence, uniqueness and numerical stability of the representation depend on the choice of parameters L and L' . For numerically stable representations, L and L' must satisfy $LL' \leq N$, or equivalently that $L \leq K$. The case where $L=K$, is called the critical sampling, and the case $L < K$ is called the over-sampling. The synthesis window $\tilde{g}(n)$ is a periodic extension (by N) of $g(n)$ which is normalized to unit energy for definiteness [3].

In general, the set of time and frequency shifted window functions, i.e., Gabor logons, $\{\tilde{g}_{m,k}(n)\}$ forms a non-orthogonal basis for the square-summable sequences space $\ell^2(\mathfrak{R})$. Hence the calculation of the Gabor coefficients is not a

simple task since projection by the usual inner product cannot be used. One of the methods [3], uses an auxiliary function $\mathbf{g}(n)$ called the biorthogonal window or dual function of. $g(n)$ Then the Gabor coefficients $\{c_{m,k}\}$ can be evaluated by

$$c_{m,k} = \sum_{n=0}^{N-1} x(n) \tilde{\mathbf{g}}_{m,k}^*(n) \quad (3)$$

where the analysis functions are

$$\tilde{\mathbf{g}}_{m,k}(n) = \tilde{\mathbf{g}}(n - mL) e^{jw_k n} \quad (4)$$

where again $\tilde{\mathbf{g}}(n)$ is a periodic version of the dual window $\mathbf{g}(n)$. Completeness condition of the basis set is obtained by substituting (3) into (1) to get that

$$\sum_{m=0}^{M-1} \sum_{k=0}^{K-1} \tilde{\mathbf{g}}_{m,k}(n) \tilde{\mathbf{g}}_{m,k}^*(\ell) = \mathbf{d}(n - \ell) \quad (5)$$

where $\mathbf{d}(\cdot)$ denotes the Dirac delta function. The above completeness relation yields equivalent but simpler bi-orthogonality condition between the analysis and synthesis basis sets via the discrete Poisson-sum formula [3]:

$$\sum_{n=0}^{N-1} \tilde{\mathbf{g}}(n + mK) e^{-j\frac{2\pi}{L}kn} \tilde{\mathbf{g}}^*(n) = \frac{L}{K} \mathbf{d}_m \mathbf{d}_k \quad (6)$$

for $0 \leq m \leq L-1$, $0 \leq k \leq L-1$. The analysis window $\mathbf{g}(n)$ is obtained by solving the equation system of the above biorthogonality condition.

Gabor analysis basis $\{\tilde{\mathbf{g}}_{m,k}(n)\}$ with a fixed window and sinusoidal modulation tiles the time-frequency plane in a rectangular fashion causing a constant bandwidth analysis. Constant bandwidth methods, such as spectrogram [1] and the Gabor expansion provide signal representations with poor time-frequency resolution [4]. Recently, representations on a non-rectangular TF grid has attracted a considerable attention [6],[10]. A non-rectangular lattice is more appropriate for the TF analysis of signals with time-varying frequency content. Thus the motivation for a fractional Gabor analysis.

3. A FRACTIONAL GABOR EXPANSION

We define a discrete fractional Gabor expansion for $x(n), 0 \leq n \leq N-1$, as follows:

$$x(n) = \sum_{m=0}^{M-1} \sum_{k=0}^{K-1} c_{m,k,a} \tilde{g}_{m,k,a}(n) \quad (7)$$

where $c_{m,k,a}$ are the fractional Gabor coefficients, a is the order of the fraction, and the basis functions are

$$\tilde{g}_{m,k,a}(n) = \tilde{g}(n - mL)W_{a,k}(n)$$

Here $\tilde{g}(n)$ is a periodic version of a unit energy Gabor window $g(n)$ and $W_{a,k}(n)$ is the fractional kernel,

$$W_{a,k}(n) = e^{j \left[-\frac{1}{2}(n^2 + (\mathbf{w}_k \sin a)^2) \cot a + \mathbf{w}_k n \right]}$$

where $\mathbf{w}_k = 2\mathbf{p}k / K$. The kernel above is similar to the Fractional Fourier Series basis functions [11]. The expansion in (7) reduces to the traditional Gabor for $a = \delta/2$. The parameters M , K , L , and L' , are same as in the traditional Gabor expansion. In our derivations, we always consider the general, oversampled case, i.e., $L < K$. The Gabor coefficients can be evaluated as before by

$$c_{m,k,a} = \sum_{n=0}^{N-1} x(n) \tilde{g}_{m,k,a}^*(n) \quad (8)$$

where the analysis functions are

$$\tilde{g}_{m,k,a}(n) = \tilde{g}(n - mL)W_{a,k}(n)$$

and $\tilde{g}(n)$ is periodic version of a $g(n)$ that is solved from a fractional biorthogonality condition between $g(n)$ and $\mathbf{g}(n)$

The completeness condition for the fractional Gabor basis, is obtained by substituting (8) in (7),

$$\begin{aligned} x(n) &= \sum_{m=0}^{M-1} \sum_{k=0}^{K-1} \left(\sum_{\ell=0}^{N-1} x(\ell) \tilde{g}^*(\ell - mL) W_{a,k}^*(\ell) \right) \\ &\quad \times \tilde{g}(n - mL) W_{a,k}(n) \\ &= \sum_{l=0}^{N-1} \sum_{m=0}^{M-1} \sum_{k=0}^{K-1} \tilde{g}(n - mL) \tilde{g}^*(\ell - mL) \\ &\quad \times e^{j \left[-\frac{1}{2}(n^2 - \ell^2) \cot a + \mathbf{w}_k (n - \ell) \right]} \end{aligned}$$

Then we obtain that the windows must satisfy the following completeness relation:

$$\begin{aligned} \sum_{m=0}^{M-1} \sum_{k=0}^{K-1} \tilde{g}(n - mL) \tilde{g}^*(\ell - mL) e^{j \left[-\frac{1}{2}(n^2 - \ell^2) \cot a \right]} \\ \times e^{j \mathbf{w}_k (n - \ell)} = \mathbf{d}(n - \ell) \end{aligned} \quad (9)$$

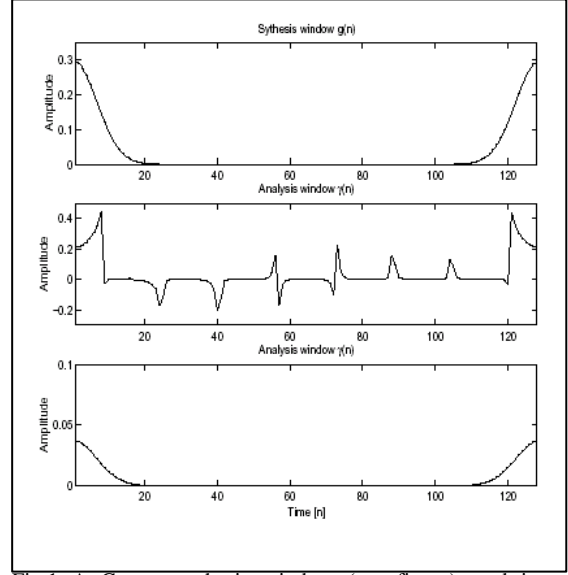


Fig.1 A Gauss synthesis window (top figure), and its biorthogonal windows in critical (middle) and oversampling (bottom) cases.

The fractional biorthogonality condition that we need to solve the analysis or dual function $\mathbf{g}(n)$ is obtained from the above completeness relation using discrete Poisson sum formula as

$$\begin{aligned} \sum_{n=0}^{N-1} \tilde{g}^*(n + mK) e^{j k \frac{2\mathbf{p}}{L} (n + mK)} \tilde{g}(n) \\ \times e^{j \left(nmK + \frac{m^2 K^2}{2} \right) \cot a} = \frac{L}{K} \mathbf{d}_m \mathbf{d}_k \end{aligned}$$

$$0 \leq m < L-1, \quad 0 \leq k < L-1$$

Completeness and biorthogonality conditions given in equations (9) and (10) reduce to the conditions in the traditional case [3] for $a = \delta/2$. This indicates that the above fractional expansion is a generalization of the discrete Gabor expansion. In Fig. 1, we show a Gauss window $g(n)$, $n=0,1,\dots,127$ on the top figure, and its biorthogonal $\mathbf{g}(n)$ for two different set of sampling parameters obtained by solving equation (10) with $a = \delta/4$. The window in the middle is obtained using $L=16$, $K=16$ that is the critical sampling and the window at the bottom is calculated with $L=8$, $K=64$ as an example of the oversampling.

4. SIMULATION RESULTS

We consider a signal composed of two linear chirps. Using our fractional Gabor method, we

analyzed the signal with two different fractional orders. Figs. 2 and 3 show the magnitude squared fractional Gabor coefficients, $|C_{m,k,\hat{a}}|^2$, of this two-chirp signal with $\hat{a}=\delta/4$ and $\hat{a}=3\delta/8$ respectively. Notice that, the component that is matched by the analysis angle becomes a narrow-band signal and represented with higher resolution.

5. CONCLUSIONS

In this paper, we present a discrete fractional Gabor expansion on a flexible, non--rectangular TF plane for the analysis of non--stationary signals. We give the completeness and biorthogonality conditions of this new expansion. Simulations show that the fractional expansion gives high resolution representations.

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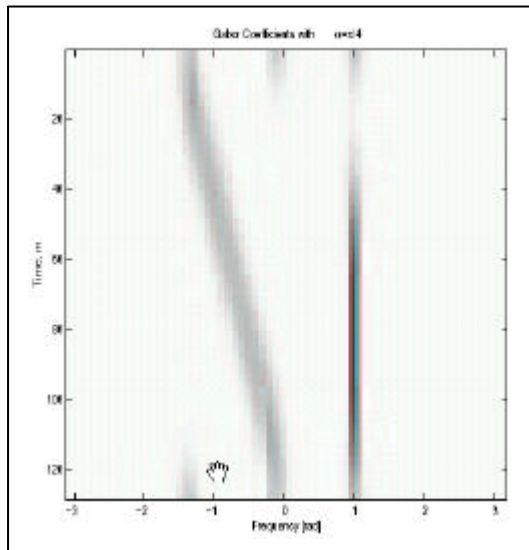


Fig.2. Gabor coefficients of the two-chirp signal using fractional order $\alpha = \delta/4$

